

IN THE CLAIMS

Please amend claims 1-13 as follows:

1. (Currently Amended) A parametric encoder ~~(100, 100')~~ for encoding an audio or speech ~~signal s~~ signal into sinusoidal code data, comprising:

- a segmentation unit ~~(110, 110')~~ for segmenting said ~~signal s~~ signal into at least one segment  ~~$x(n)$~~ ;

- a calculation unit ~~(120, 120')~~ for calculating said sinusoidal code data in the form of the phase and amplitude data of an ~~given~~ extension  ~~$\hat{x}(n)$~~  from the segment  ~~$x(n)$~~  such that the extension  ~~$\hat{x}(n)$~~  approximates the segment  ~~$x(n)$~~  as good as possible for a given criterion;

~~characterised in that~~ wherein

the calculation unit ~~(120, 120')~~ is adapted to calculate the sinusoidal code data  $\theta_k^i, d_j^i$  and  $e_j^i$  for the ~~following extension  $\hat{x}$~~  extension represented by:

$$\hat{x} = \sum_{i=1}^L Ci = \sum_{i=1}^L \sum_{j=0}^{J-1} [d_j^i f_j(n) \cos(\Theta^i(n)) + e_j^i f_j(n) \sin(\Theta^i(n))]$$

with

$$\Theta^i(n) = \sum_{k=1}^{K-1} \theta_k^i n^k$$

wherein:

i, j, k, L, J, K : represent parameters;

n : represents a discrete time parameter;

Ci : represents the i'th component of the ~~extension~~ extension;

$\theta_k^i$  : represents the phase coefficient as one of said sinusoidal data;

$f_j$  : represents the jth instance out of the set of J linearly independent functions;

$\Theta^i$  : is a phase; and

$d_j^i, e_j^i$  : represent the linearly involved amplitude values of the components representing parts of said sinusoidal data.

2. (Currently Amended) The parametric encoder according to claim 1, ~~characterised in that~~ wherein  $f_j(n) = n^j$ .

3. (Currently Amended) The parametric encoder according to claim 1, ~~characterised in that~~ wherein the calculation unit (120) comprises:

- a frequency estimation unit ~~(122)~~ for determining a plurality of  $L \times K$  phase coefficients  $\theta_k^i$  with  $i=1-L$  and  $k=1-K$  for all components  $C_i$  of the extension  $\hat{x}(n)$  representing the received segment  $x(n)$ ;

- a pattern generating unit ~~(124)~~ for calculating a plurality of  $L$  phases  $\Theta^i(n)$  with  $i=1-L$  from the phase coefficients  $\theta_k^i$  according to:

$$\Theta^i(n) = \sum_{k=1}^{K-1} \theta_k^i n^k$$

and for generating a plurality of  $J \times L$  pairs of patterns  $p_{ij}^1, p_{ij}^2$  for the components  $C_i$  with  $i=1-L$  according to:

$$p_{ij}^1 = f_j(n) \cos(\Theta^i(n)) \text{ and } p_{ij}^2 = f_j(n) \sin(\Theta^i(n))$$

- for  $i = 1-L$  and  $j = 0-(J-1)$ ; and
- an amplitude estimation unit ~~(126)~~ for determining a plurality of JxL amplitudes  $d_j^i$  for the patterns  $p_{ij}^1$  and a plurality of JxL amplitudes  $e_j^i$  for the patterns  $p_{ij}^2$  of all components  $C_i$  of ~~the extension  $\hat{x}$~~  extension;
- wherein the sinusoidal data  $\theta_k^i$ ,  $d_j^i$  and  $e_j^i$  is at least approximately ~~optimised~~ optimized for ~~the a~~ a criterion that the weighted squared error E between the ~~segment x~~ segment and its ~~extension  $\hat{x}$  is minimised~~ extension is minimized.

4. (Currently Amended) The parametric encoder according to claim 1, ~~characterised by further comprising a multiplexer (130)~~ for merging said sinusoidal code data into a data stream.

5. (Currently Amended) The parametric encoder according to claim 1, ~~characterised in that wherein~~ wherein the calculation unit ~~(120')~~ comprises:

- a frequency estimation unit ~~(122')~~ for determining a plurality of K phase coefficients  $\theta_k^i$  with  $k=1-K$  for the component  $C_i$  from an input value  $\varepsilon_{i-1}$ ; wherein for the first component  $C_1$  with  $i=1$  the input value is set to  $\varepsilon_0 = x(n)$ , where the segment is  $x(n)$ ;

- a pattern generating unit ~~(124')~~ for calculating the phases  $\Theta^i$  for the component  $C_i$  from said plurality of phase coefficients  $\theta_k^i$  according to:

$$\Theta^i(n) = \sum_{k=1}^K \theta_k^i n^k$$

and for generating a plurality of  $2 \times J$  patterns  $p_{ij}^1, p_{ij}^2$  with  $j=1-J$  for the component  $C_i$  with:

$$p_{ij}^1 = j(n) \cos(\Theta^i(n)) \text{ and } p_{ij}^2 = fj(n) \cos(\Theta^i(n));$$

- an amplitude estimation unit ~~(126')~~ for determining a plurality of J amplitudes  $d_j^1$  and of J amplitudes  $e_j^i$  for said patterns of the component  $C_i$  from the ~~received segment  $x(n)$~~  and from the ~~received~~ plurality of  $2 \times J$  patterns  $p_{ij}^1, p_{ij}^2$ ;

- a ~~synthesiser (128')~~ synthesizer for re-constructing the component  $C_i$  from said plurality of  $2 \times J$  patterns  $p_{ij}^1$ ,  $p_{ij}^2$  and form the plurality of amplitudes  $d_j^i$  and  $e_j^i$  according to:

$$C_i = \sum_{j=0}^{J-1} [d_j^i f_j(n) \cos(\Theta^i(n)) + e_j^i f_j(n) \sin(\Theta^i(n))]$$

and

- a ~~subtraction~~ subtraction unit ~~(129')~~ for ~~subtracting~~ subtracting said component  $C_i$  from the input value  $\varepsilon_{i-1}$  in order to feed the resulting difference  $\varepsilon_i$  as new input value forward to the input of the frequency estimation unit ~~(122')~~ for calculating the sinusoidal code data representing the component  $C_{i+1}$ ;

wherein the sinusoidal data  $\theta_k^i$ ,  $d_j^i$  and  $e_j^i$  is ~~optimised~~ optimized for the ~~a~~ a criterion that the weighted squared error  $E$  between the ~~segment x segment~~ and the ~~extension  $\hat{x}$  extension~~ is ~~minimised~~ minimized.

6. (Currently Amended) A parametric coding method for encoding an audio or speech ~~signal s~~ signal into sinusoidal code data, comprising the ~~steps~~ acts of:

- segmenting the ~~signal s~~ signal into at least one segment  $x(n)$ ; and

- calculating said sinusoidal code data in the form of phase and amplitude data of an given extension  $\hat{x}$  ~~extension~~ from the segment  $x(n)$  such that the ~~extension~~  $\hat{x}$  extension approximates the segment  $x(n)$  ~~as good as possible for a given criterion,~~

~~characterised in that~~ wherein

- the ~~extension~~  $\hat{x}$  extension is defined ~~to~~ as:

$$\hat{x} = \sum_{i=1}^L C_i = \sum_{i=1}^L \sum_{j=0}^{J-1} [d_j^i f_j(n) \cos(\Theta^i(n)) + e_j^i f_j(n) \sin(\Theta^i(n))]$$

with

$$\Theta^i(n) = \sum_{k=1}^K \theta_k^i n^k$$

wherein:

$i$  : represents a component  $C_i$  of the ~~extension~~  $\hat{x}(n)$  extension;

$j, k, L, J, K$  : represent parameters;

$n$  : represents a discrete time parameter;

$f_j$  : represents the  $j$ th instance out of the set of  $J$  linearly independent functions;

$\theta_k^i$  : represents the phase coefficient as one of said sinusoidal data

$\Theta^i$  : is a phase; and

$d_j^i, e_j^i$  : represent the linearly involved amplitude values of the components representing parts of said sinusoidal data.

7. (Currently Amended) The method according to claim 6,  
~~characterised in that wherein  $f_j(n) = n^j$ .~~

8. (Currently Amended) The method according to claim 6,  
~~characterised in that wherein the frequencies phase coefficients  $\theta_i^j$~~   
are defined by picking peak frequencies in the frequency domain of  
~~the extension  $\hat{x}$  extension.~~



9. (Currently Amended) The method according to claim 6,  
~~characterised in that~~ wherein, for fulfilling ~~the~~ a criterion that  
the weighted squared error between the ~~segment x~~ segment and the  
~~extension  $\hat{x}$~~  extension is minimized, the definition of the optimal  
amplitudes  $d_j^i$  and  $e_j^i$  comprises the ~~steps~~ acts of:

- determining a plurality of  $L \times K$  phase coefficients  $\theta_k^i$  with  
 $i=1-L$  and  $k=1-K$  for all components  $C_i$  of the ~~received segment  $x(n)$~~   
segment;

- calculating a plurality of  $L$  phases  $\Theta^i(n)$  with  $i=1-L$  from  
the phase coefficients  $\theta_k^i$  according to:

$$\Theta^i(n) = \sum_{k=1}^K \theta_k^i n^k ;$$

- generating a plurality of  $J \times L$  pairs of patterns  $p_{ij}^1, p_{ij}^2$   
for the components  $C_i$  with  $i=1-L$  according to:

$$p_{ij}^1 = f_j(n) \cos(\Theta^i(n)) \text{ and } p_{ij}^2 = f_j(n) \sin(\Theta^i(n)); \text{ and}$$

- determining a plurality of JxL amplitudes  $d_j^i$  and a plurality of JxL amplitudes  $e_j^i$  for all the pairs of patterns  $p_{ij}^1$ ,  $p_{ij}^2$  of all components  $C_i$  of the extension  $\hat{x}$ .

10. (Currently Amended) The method according to claim 6, ~~characterised in that wherein, for fulfilling the a criterion that the weighted squared error between the segment x segment and the extension  $\hat{x}$  extension is minimized, the a definition of the amplitudes  $d_j^i$  and  $e_j^i$  comprises the steps acts of:~~

- a) setting  $i = 1$
- b)  $\varepsilon_{i-1} = \varepsilon_0 = x(n)$ ;
- c) determining a plurality of K phase coefficients  $\theta_k^i$  with  $k=1-K$  for the component  $C_i$  from an input value  $\varepsilon_{i-1}$ ;
- d) calculating the phases  $\Theta^i$  for the component  $C_i$  from said plurality of phase coefficients  $\theta_k^i$  according to:

$$\Theta^i(n) = \sum_{k=1}^K \theta_k^i n^k$$

- e) generating a plurality of 2xJ patterns  $p_{ij}^1$ ,  $p_{ij}^2$  with

$j=0-(J-1)$  for the component  $C_i$  with:

$$p_{ij}^1 = f_j(n) \cos(\Theta^i(n)) \text{ and } p_{ij}^2 = f_j(n) \sin(\Theta^i(n));$$

f) determining a plurality of  $J$  amplitudes  $d_j^i$  and of  $J$  amplitudes  $e_j^i$  for said patterns for the component  $C_i$  from the ~~received segment  $x(n)$~~  segment and from the ~~received plurality of  $2 \times J$  patterns~~  $p_{ij}^1, p_{ij}^2$ ;

g) constructing the component  $C_i$  from said plurality of  $J$  pairs of patterns  $p_{ij}$  and from the plurality of amplitudes  $d_j^i$  and  $e_j^i$  according to:

$$C_i = \sum_{j=0}^{J-1} [d_j^i f_j(n) \cos(\Theta^i(n)) + e_j^i f_j(n) \sin(\Theta^i(n))]$$

h) ~~subtracting~~ subtracting said component  $C_i$  from the input value  $\varepsilon_{i-1}$  in order to calculate a resulting difference  $\varepsilon_i$ ;

i) checking if  $i \geq L$  wherein  $L$  represents a given number ~~of~~ number of components ;

j) if  $i < L$  repeat the method ~~steps~~ acts by starting again from ~~step~~ act c) with  $i = i+1$ ; and

k) if  $i \geq L$  the sinusoidal code data of all  $L$  components of the ~~extension~~  $\hat{x}$ -extension have been calculated and thus the ~~process has finished.~~

11. (Currently Amended) A parametric decoder ~~(400)~~ for re-constructing an ~~approximation~~  $\hat{s}$ -approximation of an audio or speech ~~signal  $s$  from transmitted~~ signal from transmitted or restored code data, comprising:

- a selecting unit ~~(420)~~ for selecting sinusoidal code data representing ~~segments~~  $\hat{x}$ -segments of the ~~approximation~~  $\hat{s}$ -approximation from said ~~received-transmitted or restored code data;~~
- a ~~synthesiser~~ (440) synthesizer for re-constructing said ~~segments~~  $\hat{x}$ -segments from said ~~received-sinusoidal code data; and~~
- a joining unit ~~(460)~~ for joining consecutive ~~segments~~  $\hat{x}$ -segments to form said ~~approximation~~  $\hat{s}$ -approximation of the audio or speech ~~signal  $s$  signal;~~

wherein the sinusoidal code data is a plurality of frequency and amplitude values for at least one component of said ~~segment~~  $\hat{x}$  segments;

~~characterised in that~~ wherein

- the ~~synthesiser~~ synthesizer is adapted to re-construct said ~~segments~~  $\hat{x}$  segments from said sinusoidal code data according to an extension represented by the following formula:

$$\hat{x} = \sum_{i=1}^L C_i = \sum_{i=1}^L \sum_{j=0}^{J-1} [d_j^i f_j(n) \cos(\Theta^i(n)) + e_j^i f_j(n) \sin(\Theta^i(n))]$$

with

$$\Theta^i(n) = \sum_{k=1}^K \theta_k^i n^k$$

wherein:

i : represents a component  $C_i$  of the ~~extension~~  $\hat{x}(n)$  extension;

j, k, L, J, K : represent parameters;

n : represents a discrete time parameter;

$f_j$  : represents the jth instance out of the set of J linearly independent functions;

$\theta_k^i$  : represents the phase coefficient value as one of said sinusoidal data

$\Theta^i$  : is a phase; and

$d_j^i, e_j^i$  : represent the linearly involved amplitude values of the components representing parts of said sinusoidal data.

12. (Currently Amended) Decoding method for reconstructing an ~~approximation- $\hat{s}$ -approximation~~ of an audio or speech ~~signal-s~~ signal from transmitted or restored code data, comprising the ~~steps~~ acts of selecting sinusoidal code data representing ~~segments- $\hat{x}$~~  segments of the ~~approximation- $\hat{s}$ -approximation~~ from said received transmitted or restored code data;

- re-constructing said ~~segments- $\hat{x}$~~  segments from said received sinusoidal code data; and
- joining consecutive ~~segments- $\hat{x}$~~  ones of said segments together in order to form said ~~approximation- $\hat{s}$~~  of the audio or speech ~~signal-s~~ signal;

- wherein the sinusoidal code data is a plurality of phase and amplitude values for at least one component of said ~~segment  $\hat{x}$~~  segment,

~~characterised in that wherein~~

- in said re-construction step act, the ~~segments  $\hat{x}$~~  segments are re-constructed from said sinusoidal code data according to an extension represented by the following formula:

$$\hat{x} = \sum_{i=1}^L C_i = \sum_{i=1}^L \sum_{j=0}^{J-1} [d_j^i f_j(n) \cos(\Theta^i(n)) + e_j^i f_j(n) \sin(\Theta^i(n))]$$

with

$$\Theta^i(n) = \sum_{k=1}^K \theta_k^i n^k$$

wherein:

i : represents a component  $C_i$  of the ~~extension  $\hat{x}(n)$~~  extension;

j, k, L, J, K : represent parameters;

n : represents a discrete time parameter;

- $f_j$  : represents the  $j$ th instance out of the set of  $J$  linearly independent functions;
- $\theta_k^i$  : represents the phase coefficient as one of said sinusoidal data
- $\Theta^i$  : is a phase; and
- $d_j^i, e_j^i$  : represent the linearly involved amplitude values of the components representing parts of said sinusoidal data.

13. (Currently Amended) Data stream comprising sinusoidal code data representing segments  ~~$\hat{x}$~~  a segment of an approximation  $\hat{x}$  ~~approximation~~ of an audio or speech signal, wherein the sinusoidal code data is a plurality of phase and amplitude values for at least one component of said ~~segment  $\hat{x}$~~  segment, characterised in that wherein the segment  $\hat{x}$  segment is defined according to an extension represented by to:

$$\hat{x} = \sum_{i=1}^L C_i = \sum_{i=1}^L \sum_{j=0}^{J-1} \left[ d_j^i f_j(n) \cos(\Theta^i(n)) + e_j^i f_j(n) \sin(\Theta^i(n)) \right]$$



with

$$\Theta^i(n) = \sum_{k=1}^K \theta_k^i n^k$$

wherein:

i	:	represents a component C <sub>i</sub> of the <del>extension</del> $\hat{x}(n)$ <del>extension</del> ;
j, k, L, J, K	:	represent parameters;
n	:	represents a discrete time parameter;
f <sub>j</sub>	:	represents the jth instance out of the set of J linearly independent functions;
$\theta_k^i$	:	represents the phase coefficient as one of said sinusoidal data
$\Theta^i$	:	is a phase; and
$d_j^i, e_j^i$	:	represent the linearly involved amplitude values of the components representing parts of said sinusoidal data.

14. (Original) Storage medium on which a data stream as claimed in claim 13 has been stored.